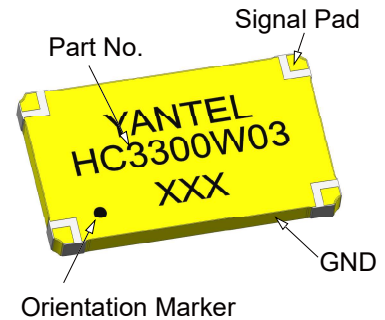


#### Description

High-power broadband surface-mounted and embedded coupler series, realizing the power synthesis and distribution of microwave high-power amplifier system, signal acquisition and other functions. Used in active phased array radar, microwave transceiver components, microwave amplifiers, radio stations, satellite communications and other projects, to provide standardized and customized high-quality and reliable products.

The performance and reliability indexes are in line with international products, and the pin definition and package size are compatible with international products, realizing 100% in-situ replacement.



#### Features:

- 600-6000MHz
- High Power
- Very Low Loss
- Tight Amplitude Balance
- High Isolation
- Low VSWR
- Good Repeatability
- CTE compatible with FR4, G-10, RF-35, RO4350B and polyimide
- Immersion gold, prevent surface oxidation & scratch
- RoHS Compliant

#### Electrical Specifications

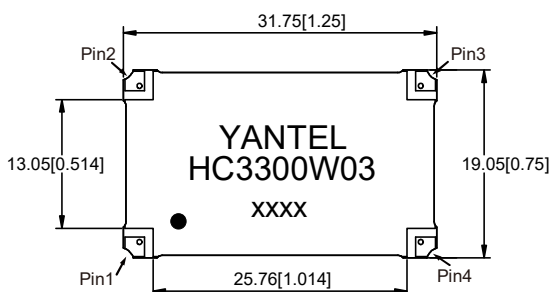
Frequency	Isolation	Insertion Loss	Return Loss	Amplitude Balance
<i>MHz</i>	<i>dB Min</i>	<i>dB Max</i>	<i>dB</i>	<i>dB Max</i>
600 - 6000	18	1.0	14	0.9
<b>Phase Balance</b>	<b>Power</b>	<b>Operating Temp.</b>		
<i>Degrees</i>	<i>Avg. CW Watts</i>	<i>°C</i>		
90 ± 6.0	100	-55 to +95		

#### Notes:

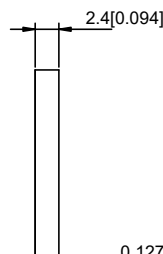
1. All the above data are based on specified demo board.
2. Insertion loss: Thru board loss has been removed.

#### Mechanical Outline

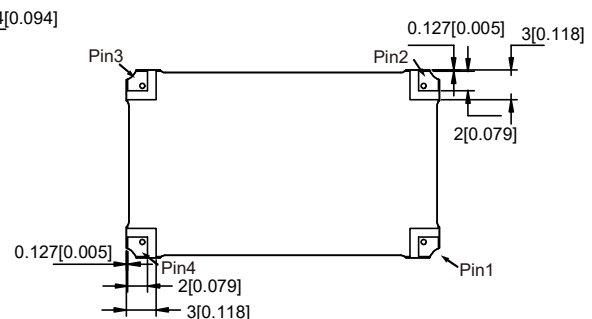
##### TOP VIEW



##### SIDE VIEW



##### BOTTOM VIEW



- Note:**
1. NC represents no ground plane permitted in hatched area
  2. All Dimensions show in mm[inch]
  3. RoHS Compliant in accordance with EU Directive(2011/65/EU)
  4. REACH Compliant in accordance with Regulation (EC) No 1907/2006
  5. Dimension tolerance: ±0.20mm

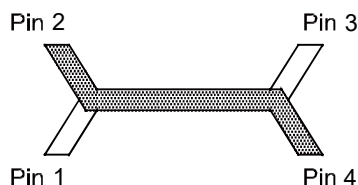
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## Hybrid Coupler Pin Configuration

The HC3300W03 has an orientation marker to denote Pin 1. Once port one has been identified the other ports are known automatically. Please see the chart below for clarification:



Configuration	Pin 1	Pin 2	Pin 3	Pin 4
<b>Splitter</b>	Input	-3dB $\angle \theta$	-3dB $\angle \theta - 90$	Isolated
<b>Splitter</b>	-3dB $\angle \theta$	Input	Isolated	-3dB $\angle \theta - 90$
<b>Splitter</b>	-3dB $\angle \theta - 90$	Isolated	Input	-3dB $\angle \theta$
<b>Splitter</b>	Isolated	-3dB $\angle \theta - 90$	-3dB $\angle \theta$	Input
<b>*Combiner</b>	$A \angle \theta - 90$	Output	Isolated	$A \angle \theta$
<b>*Combiner</b>	Output	$A \angle \theta - 90$	$A \angle \theta$	Isolated
<b>*Combiner</b>	Isolated	$A \angle \theta$	$A \angle \theta - 90$	Output
<b>*Combiner</b>	$A \angle \theta$	Isolated	Output	$A \angle \theta - 90$

\*Note: "A" is the amplitude of the applied signals. When two quadrature signals with equal amplitudes are applied to the coupler as described in the table, they will combine at the output port. If the amplitudes are not equal, some of the applied energy will be directed to the isolated port.

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#### Definition of Measured Specifications

Parameter	Definition	Mathematical Representation
<b>VSWR</b> (Voltage Standing Wave Ratio)	The impedance match of the coupler to a 50Ω system. A VSWR of 1:1 is optimal.	$VSWR = \frac{V_{max}}{V_{min}}$ Vmax = voltage maxima of a standing wave Vmin = voltage minima of a standing wave
<b>Return Loss</b>	The impedance match of the coupler to a 50Ω system. Return Loss is an alternate means to express VSWR.	$\text{Return Loss (dB)} = 20 \log \frac{VSWR + 1}{VSWR - 1}$
<b>Insertion Loss</b>	The input power divided by the sum of the power at the two output ports.	$\text{Insertion Loss (dB)} = 10 \log \frac{P_{in}}{P_{cpl} + P_{transmission}}$
<b>Isolation</b>	The input power divided by the power at the isolated port.	$\text{Isolation (dB)} = 10 \log \frac{P_{in}}{P_{iso}}$
<b>Phase Balance</b>	The difference in phase angle between the two output ports.	Phase at coupled port – Phase at transmission port
<b>Amplitude Balance</b>	The power at each output divided by the average power of the two outputs.	$10 \log \left( \frac{P_{cpl}}{P_{cpl} + P_{transmission}} \right) \text{ or } 10 \log \left( \frac{P_{transmission}}{P_{cpl} + P_{transmission}} \right)$

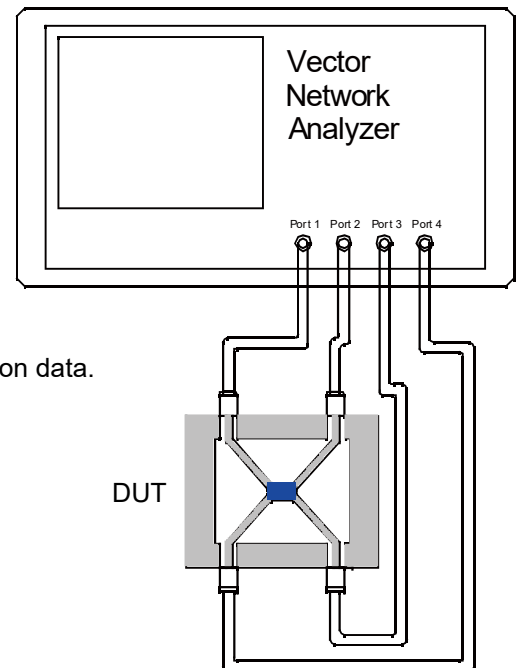
#### Test Method

1. Calibrating your vector network analyzer.
2. Connect the VNA 4 Port to DUT respectively.
3. Measure the data of coupling through port 1 to port 4(S41).
4. Measure the data of transmission through port 1 to port 3(S31).
5. Measure the data of isolation through port 1 to port 2(S21).
6. Measure the data of phase port 4 & port 3(port 1 feeding).
7. Measure the data of return loss port 1, port 2, port 3 & port 4.
8. According to the above data to calculate insertion loss, amplitude balance & phase.

#### Note:

1. When calculating insertion loss at room temperature, demo board loss should be removed from both coupling & transmission data.
- Please refer to the below table for demo board loss :

Frequency Range(MHz)	Demo Board Loss (dB) @25°C
<b>470-860</b>	0.07
<b>800-1000</b>	0.10
<b>1200-1700</b>	0.15
<b>1700-2000</b>	0.15
<b>2000-2300</b>	0.20
<b>2300-2700</b>	0.25

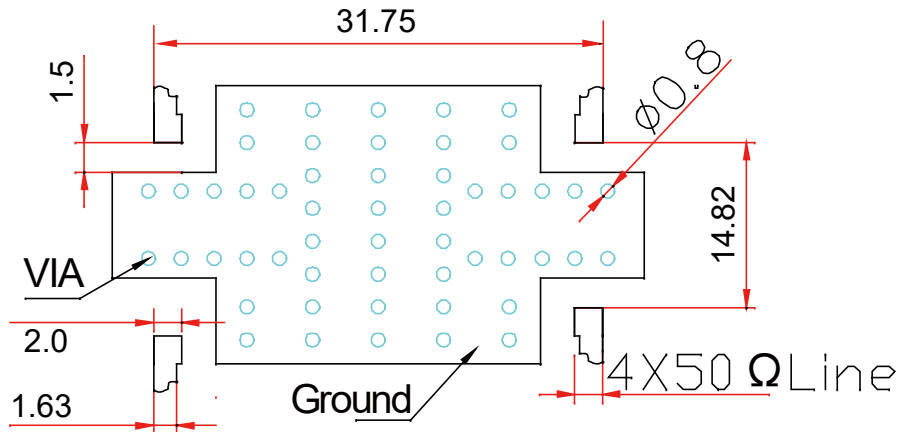


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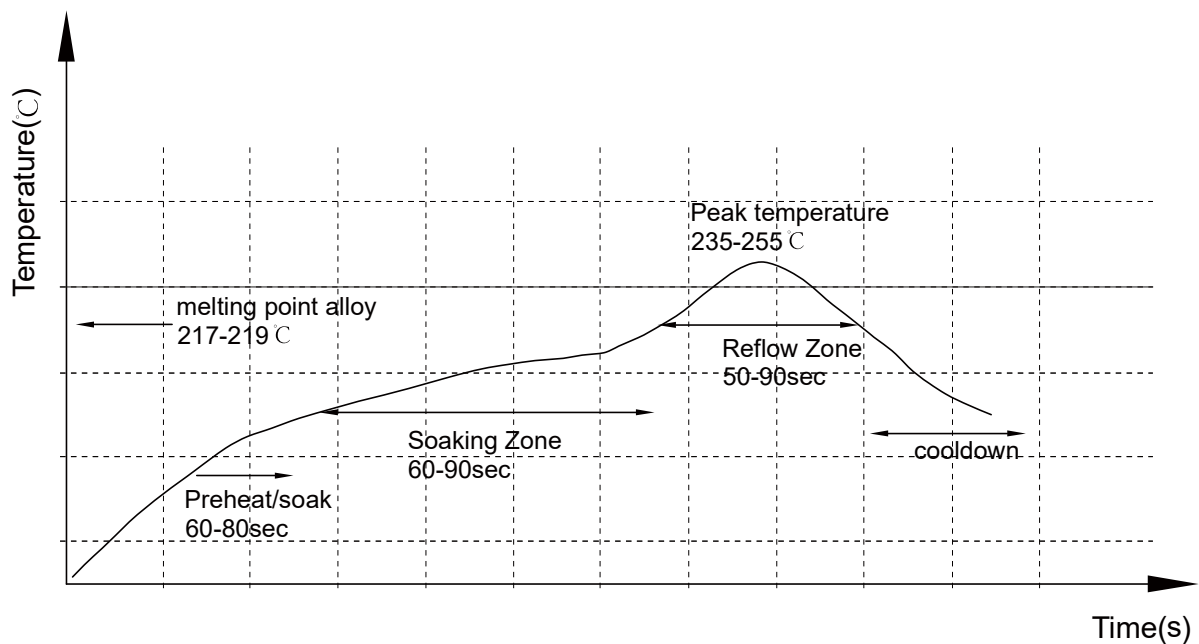
## Recommended PCB Layout



### NOTE:

1. 50 $\Omega$  line width is shown above designing from RO4350B dielectric thickness 0.762mm; copper 1 OZ
2. Bottom side of the PCB is continuous ground plane.
3. All dimensions shown in mm.

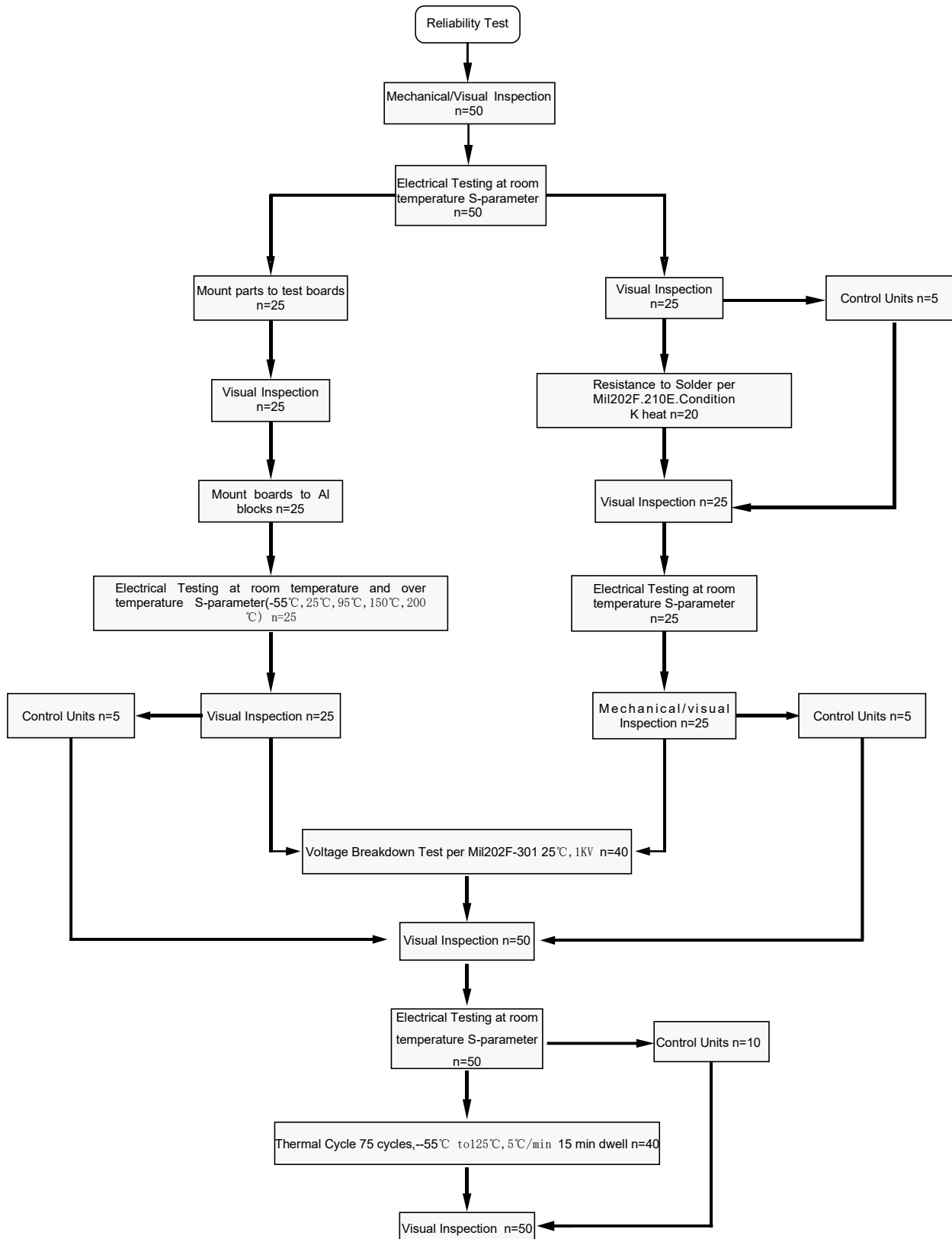
## Reflow Profile



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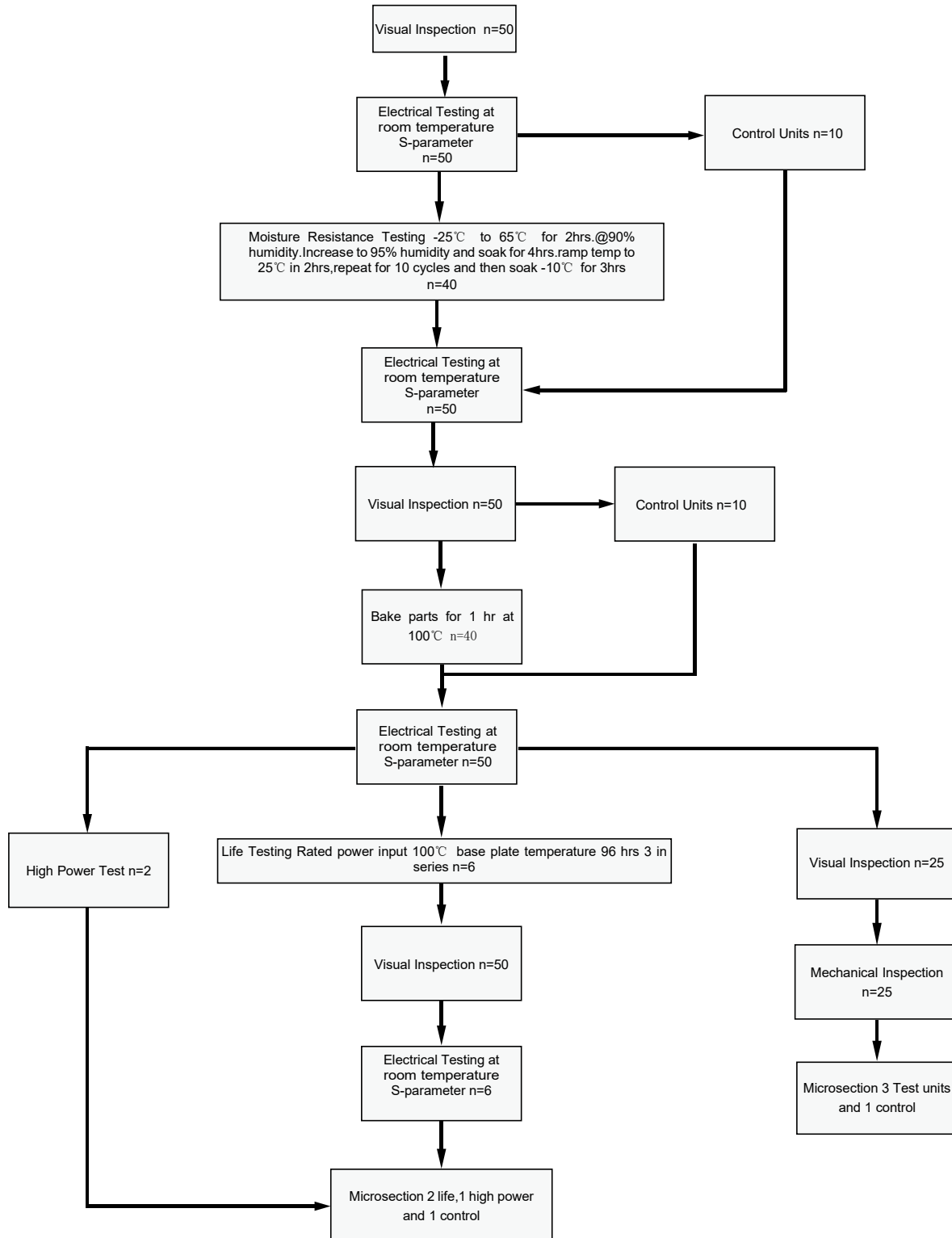
#### Reliability Test Flow



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#### Reliability Test Flow



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